

Appendix B. Program Listing

```
/*
*****
*
* Program to control LORAN-C radio
*
* This program controls a special-purpose radio designed to receive
* transmissions from the US Coast Guard LORAN-C navigation system.
* These stations operate on an assigned radio frequency of 100 kHz
* and can be received over the continental US, adjacent coastal areas
* and significant areas elsewhere in the world.
*
* The radio, which is contained in a separate, shielded box, receives
* the signals, which consist of an eight-pulse biphas-modulated
* pulse group transmitted at a 1-kHz rate. Each of these pulse groups
* is repeated at an interval characteristic of the particular LORAN-C
* chain, which consists of a master station and up to three slave
* stations. The radio includes a synchronous detector driven by a
* quadrature-phase clock, two integrators with adjustable gain and
* an signal-level detector used to derive the agc voltage.
*
* The radio is controlled by this program using a special-purpose
* interface, which converts the receiver signals using an
* analog/digital converter and multiplexor, and generates the digital
* timing and analog control signals using an AMD 9513A System Timing
* Controller (STC) chip, two digital/analog converters and
* miscellaneous logic components. The radio provides three analog
* signals, one for the in-phase integrator, another for the
* quadrature-phase integrator and a third for the agc. This program
* computes the master oscillator frequency adjustment and receiver
* agc voltage.
*
* The reciever includes a precision, oven-controlled crystal
* oscillator used to derive all timing signals used by the receiver
* and this program. The 5-MHz output of this oscillator is adjusted
* over a small range by this program to coincide with the LORAN-C
* signal as broadcast to within a few parts in 10e10 and is suitable
* for use as a laboratory frequency standard. The oscillator itself
* should have good intrinsic stability and setability to within less
* than 2.5 Hz at 5 MHz (0.5 ppm), since it must maintain the master
* clock to within 100 us over the pulse-code scan interval of several
* minutes.
*
* The PC running this program generates the control signals necessary
* to run the radio and produces a 1-pps signal synchronized to
* UTC(LORAN) to within a fraction of a microsecond. When manually
* adjusted using time-of-coincidence (TOC) data published by US Naval
* Observatory, this signal is suitable for use as a precision source
* of standard time. The system can generate all sorts of external
* signals as well, as programmed in the 9513A.
*
* David L. Mills (mills@udel.edu) 27 March 1992
*
*****
*
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* Current LORAN-C chains by gri (master listed first)
*
* 9990 North Pacific      St. Paul Island, Attu, Port Clarence,
*                          Narrow Cape
* 9980 Icelandic Sea
* 9970 North West Pacific Iwo Jima, Marcus Island, Hokkaido, Gesashi,
*                          Yap Island
* 9960 North East US     Seneca, Caribou, Nantucket, Carolina Beach,
*                          Dana
* 9940 West Central US   Fallon, George, Middletown, Searchlight
* 9610 South Central US
* 8970 Great Lakes US    Dana, Malone, Seneca, Baudette
* 8290 North Central US
* 7990 Mediterranean Sea Sellia Marina, Lampedusa, Kargabarun,
*                          Estartit
* 7980 South East US     Malone, Grangeville, Raymondville, Jupiter,
*                          Carolina Beach
* 7970 Norwegian Sea     Ejde, Sylt, B0, Sandur, Jan Mayen
* 7960 Gulf of Alaska     Tok, Narrow Cape, Shoal Cove
* 7930 Labrador Sea      Angissoq, Sandur, Ejde, Cape Race
* 5990 West Central Canada Williams Lake, Shoal Cove, Port Hardy
* 5930 East Central Canada Caribou, Nantucket, Cape Race
* 4990 Central Pacific    Johnson Island, Upolu Point, Kure Island
*/

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#include <stdio.h>
#include <ctype.h>
#include <bios.h>
#include <math.h>
#include <conio.h>
#include <string.h>

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/*
* Sizes of things. The pulse-group filter is a shift register with one
* stage for each 100-us bin in a 1000-us sample window plus two stages
* for a noise gate. The envelope filters consist of one stage for each
* 10-us cycle in the 300-us envelope pulse gate, for a total of 30
* samples.
*/
#define NRMS 10                /* size of pulse-group filter */
#define NENV 30               /* size of envelope filter */

/*
* Program characteristics. The field and display guard times are the
* maximum latency for the program to process samples at the end of a
* gri and for the output routines to print a line, respectively. If the
* time to the next gri is less than either of these, the next frame is
* skipped. The watchdog timeout is the maximum number of frames before
* the program abandons cycle search and reverts to pulse-group search.
* The agc averaging factor sets the time constant of integration for
* the receiver signal- level indicator. The remaining parameters
* establish the minimum and maximum median filter size and envelope and
* phase weights.
*/
#define FGUARD 1000           /* field guard time (100 us) */

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#define DGUARD 8000          /* display guard time (100 us) */
#define WATCHDOG 2000      /* watchdog timeout (frame) */
#define AGCFAC 16          /* agc averaging factor */
#define MMIN 3              /* min median filter size */
#define MMAX 10             /* max median filter size */
#define EMIN 5              /* min envelope weight */
#define EMAX 50             /* max envelope weight */
#define EFAC 1.1           /* envelope adjustment factor */
#define PMIN 5              /* min phase weight */
#define PMAX 200           /* max phase weight */
#define PFAC 1.2           /* phase adjustment factor */

/*
 * Receiver characteristics. The receiver delay is characteristic of the
 * receiver bandpass. The pulse-group offset is characteristic of the
 * pulse-group filter and noise gate.
 */
#define RCVDELAY 50        /* receiver delay (200 ns) */
#define OFFSET 21         /* pulse-group offset (10 us) */

/*
 * Receiver gain and noise gates. The vco parameter is adjusted
 * for zero nominal frequency offset. The agc parameter is adjusted for
 * nominal receiver output on the peak loran pulse of 800 mv p-p. The
 * receiver gain parameter is adjusted so that the agc threshold (knee)
 * occurs at a peak signal amplitude of 100, as determined by the status
 * display. The derived envelope factor is adjusted so the zero crossing
 * of the derived envelope signal occurs at the third cycle. The noise
 * gate parameters establish the error/false-alarm rates.
 */
#define VCO 194            /* initial vco dac */
#define AGC 162            /* initial agc dac */
#define RGAIN 2.5          /* receiver gain */
#define DERVEL 2.2908      /* derived envelope factor */
#define PGATE 3            /* pulse-group noise gate */
#define SGATE 2            /* strobe noise gate */

/*
 * The receiver agc is controlled to produce a q-channel amplitude of
 * 100, which represents a demodulator transfer function at the third
 * carrier cycle of 50 V/rad. The vco has a sensitivity of 1 Hz/V
 * reduced to 0.1 Hz/V by a pad between the dac and the vco, which
 * represents a transfer function of 0.628 rad/V-s. The dac produces 6 V
 * p-p for an input of 256V p-p, for a transfer function of .0234. The
 * ratio of the 100-kHz demodulator clock to the vco frequency (5 MHz)
 * is 1/50. The overall pll gain is the product of these factors .0147,
 * rounded up to .015 for neatness.
 */

#define VGAIN .015         /* overall loop gain */

/*
 * Timing generator definitions
 */
#define PORT 0x0300        /* controller port address */

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#define TGC PORT+0          /* stc control port (r/w) */
#define TGD PORT+1          /* stc data port (r/w) */
/*
 * Analog/digital converter definitions
 */
#define ADC PORT+2          /* adc buffer (r)/address (w) */
#define ADCGO PORT+3        /* adc status (r)/adc start (w) */
    #define START 0x01      /* converter start bit (w) */
    #define BUSY 0x01       /* converter busy bit (r) */
    #define DONE 0x80       /* conversion done bit (r) */
/*
 * Digital/analog converter definitions
 * Note: output voltage is inverted from buffer
 */
#define DACA PORT+4         /* vco (dac a) buffer (w) */
#define DACB PORT+5         /* agc (dac b) buffer (w) */
/*
 * Code generator definitions
 * Note: bits are shifted out from the lsb first
 */
#define CODE PORT+6         /* pulse-code buffer (w) */
    #define MPCA 0xCA        /* LORAN-C master pulse code group a */
    #define MPCB 0x9F        /* LORAN-C master pulse code group b */
    #define SPCA 0xF9        /* LORAN-C slave pulse code group a */
    #define SPCB 0xAC        /* LORAN-C slave pulse code group b */
/*
 * Mode register definitions
 */
#define PAR PORT+7          /* parameter buffer (w) */
    #define INTEG 0x03       /* integrator mask */
/*
 * time constant values
 *
 * 0 1.000 ms
 * 1 0.264 ms
 * 2 0.036 ms
 * 3 short caps
 */
#define GATE 0x0C           /* gate source mask */
/*
 * gate source values
 *
 * 4 always open
 * 8 group repetition interval (GRI)
 * c puldse code interval (PCI)
 * f strobe (STB)
 */
#define IEN 0x20            /* enable interrupt bit */
#define EN5 0x40           /* enable counter 5 bit */
#define ENG 0x80           /* enable gri bit */

/*
 * Timing generator (STC) commands
 */
/* argument sssss = counter numbers 5-1 */

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#define LOADDP 0x00          /* load data pointer */
    /* argument ee = element (all groups except ggg = 000 or 111) */
#define MODEREG 0x00        /* mode register */
#define LOADREG 0x08        /* load register */
#define HOLDREG 0x10        /* hold register */
#define HOLDINC 0x18        /* hold register (hold cycle increm) */
    /* argument ee = element (group ggg = 111) */
#define ALARM1 0x07         /* alarm register 1 */
#define ALARM2 0x0F         /* alarm register 2 */
#define MASTER 0x17         /* master mode register */
#define STATUS 0x1F         /* status register */
#define ARM 0x20            /* arm counters */
#define LOAD 0x40           /* load counters */
#define LOADARM 0x60        /* load and arm counters */
#define DISSAVE 0x80        /* disarm and save counters */
#define SAVE 0xA0           /* save counters */
#define DISARM 0xC0         /* disarm counters */
    /* argument nnn = counter number */
#define SETTOG 0xE8         /* set toggle output HIGH for counter */
#define CLRTOG 0xE0         /* set toggle output LOW for counter */
#define STEP 0xF0           /* step counter */
    /* argument eegg, where ee = element, ggg - counter group */
    /* no arguments */
#define ENABDPS 0xE0        /* enable data pointer sequencing */
#define ENABFOUT 0xE6       /* enable fout */
#define ENAB8 0xE7          /* enable 8-bit data bus */
#define DSABDPS 0xE8        /* disable data pointer sequencing */
#define ENAB16 0xEF         /* enable 16-bit data bus */
#define DSABFOUT 0xEE       /* disable fout */
#define ENABPFW 0xF8        /* enable prefetch for write */
#define DSABPFW 0xF9        /* disable prefetch for write */
#define RESET 0xFF          /* master reset */

/*
 * Function declarations
 */
void status(double, double, char*);

/*
 * STC setup. Note gri = 99600, pci = 300 and stb = 10 (us).
 *
 * Counter 1 generates a 200-kHz signal from the 5-MHz master VCO. This
 * signal is a slightly assymetrical square wave (duty factor 12/13). An
 * external flipflop divides this signal by two to get the 100-kHz gri
 * clock which drives counter 2. All other counters are driven from the
 * 5-MHz source. The 200-kHz and 100-kHz signals are used by the
 * synchronous demodulator in the receiver.
 *
 * Counter 2 generates the gri (pulse-code) gate, which repeats at the
 * interval assigned to the LORAN-C chain; e.g., 9960 for the Northeast
 * U.S. chain. The signal consists of a high 8-ms interval preceded by
 * a programmable low interval normally equal to the gri interval less 8
 * ms. Counter 3 generates the pulse-code (pci) gate, which enables the
 * receiver only when a pulse group is expected. The signal consists of
 * eight 300-us high intervals beginning at the high interval of counter

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* 2. Counter 4 generates the stb (cycle) gate used during envelope
* scan. The signal consists of a high 10-us interval preceded by a
* programmable interval in the range up to about 300 us.
*
* Counter 5 operates as a gated divider to drive the frequency scalar
* and output divider, which produces the output signals for external
* equipment. The gating signal is generated by counter 4, which can be
* enabled for this purpose under program control. When so enabled,
* counter 5 is stopped for the interval programmed in counter 4,
* enabling precise alignment of the frequency scalar and output divider
* to UTC. The output signal can be at 1 pps and any decimal multiple up
* to 100 kHz plus 5 MHz or, if UTC alignment is not necessary, any
* binary or decimal submultiple of 5 MHz. Note that all counters
* operate in binary mode, except the frequency scalar and output
* divider, which normally operate in bcd mode.
*/
int init[] = {
    0x0162, 12, 13, /* counter 1 (p0) */
    0x0262, 9160, 800, /* counter 2 (gri) */
    0x8162, 1500, 3500, /* counter 3 (pcx) */
    0xc162, 0, 50, /* counter 4 (stb) */
    0x0162, 25, 25 /* counter 5 (out) */
};

/*
* Standard envelope cycle amplitudes. These are matched with the
* received envelope amplitudes to compute the rms error, assuming the
* reference cycle (3) is at the reference phase in the envelope window.
* Following are the cycle amplitude values (peak-normalized to 100) in
* the standard LORAN transmission specification.
*
*          cycle    1    2    3*    4    5    6    7 */
double envcyc[7] = {4.7, 25.3, 50.8, 72.9, 88.6, 97.3,100.0};

/*
* Program variables (units)
*/
int ptrenv = 0; /* index of display cycle */
char mode = '1'; /* operating mode */
char pulse = 'm'; /* master (m) or slave (s) codes */
int par = 0x0a; /* mode register */
char kbd = ' '; /* latest keystroke */

/*
* System timing and rrelated data (units). These values provide the
* precise offset of the reference phase relative to the epoch when
* the STC chip was last reset by this program. All timing calculations
* are performed relative to this epoch. The freq and phase variables
* are computed in various places in the program, but take effect only
* at the end of the processing cycle, so that all timing calculations
* can be performed with respect to the epoch the system is actually at.
*/
int frame = 0; /* offset to reference frame (2*gri) */
int offset = 0; /* offset to reference gri (10 us) */
int strobe = 0; /* offset to reference cycle (10 us) */
int gri = 9960; /* group repetition interval (10 us) */

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int pcx = 1500;                /* pulse-code interval (200 ns) */
int freq = 0;                 /* frequency offset (10 us/frame) */
int phase = 0;               /* phase offset (10 us) */
int step = 1;                /* phase step (10 us) */
int stb[MMAX];               /* strobe median filter */
int sgate = SGATE;          /* strobe noise (max-min) */

/*
 * Various controls normally preset, but can be adjusted by keyboard
 * commands. After all, this is a prototype device.
 */
double vco = VCO;            /* vco dac signal */
double vcodac = VCO;         /* vco dac bias (dac a) */
double agc = AGC;            /* agc dac signal */
double agcdac = AGC;         /* agc dac bias (dac b) */

/*
 * Raw and processed data input from receiver. The raw data are received
 * directly from the adc and summed for both the a and b gri intervals.
 * The offset data are computed during the receiver calibration modes
 * and used to remove bias from the raw data to produce the net signal
 * valid at the end of the frame. As the envelope pointer cycles through
 * all 30 100-us cycles of the pulse gate, the i and q signal envelopes
 * are averaged separately for use in the cycle-identification and
 * phase-tracking processes. The median filters are used to suppress
 * impulse-noise and pulse-dropout. The rms error signal produced from
 * the i and q signals is used during the pulse-code identification
 * process.
 */
double isig = 0;             /* i-signal (a+b) */
double qsig = 0;             /* q-signal (a+b) */
double pmed[NENV][MMAX];    /* phase median filter */
double emed[NENV][MMAX];    /* envelope median filter */
double mgate = 0;           /* envelope median filter span */

double pcyc[NENV];           /* cycle phase */
double pfac = PMIN;          /* cycle phase weight */
double ecyc[NENV];           /* cycle amplitude */
double efac = EMIN;          /* cycle amplitude weight */
double erms[NENV];           /* cycle rms error */
double edrv[NENV];           /* cycle derived envelope */

double iofs = 307;           /* i-integrator offset */
double qofs = 308;           /* q-integrator offset */
double agcraw = 0;           /* receiver agc output (adc chan 2) */
double agcavg = 0;           /* receiver agc smoothed signal */
double agcofs = 317;         /* receiver agc offset (zero signal) */
double agcmax = 355;         /* receiver agc max signal (overload) */
double rms[NRMS];           /* pulse-group signal shift register */
double gain = RGAIN;         /* program gain */
double dervel = DERVEL;      /* derived envelope factor */
char report[257] = "\0";     /* report string for display */

/*
 * Event counters. These tally the synchronization events of interest

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    * for debugging and monitoring.
    */
int pgcnt = 0;          /* pulse-group search events */
int encnt = 0;          /* envelope search events */
int csnt = 0;           /* cycle-slip events */
int ssnt = 0;           /* strobe-slip events */
int pncnt = 0;          /* pulse-group noise events */

/*
 * Signal/noise ratios. These reveal signal quality and health of the
 * tracking processes.
 */
double psnr = 0;        /* pulse-group max-envelope/rms */
double esnr = 0;        /* envelope rms-max/rms-min */

/*
 * Main program
 */
 * Programming note: There is usually enough time between gri intervals
 * for one print statement, but not two, at least on a 286.
 */
main(argc, argv) int argc; char *argv[]; {
    int mindex = 0;          /* index of min cycle in envelope */
    int maxdex = 0;          /* index of max cycle in envelope */
    int cycle = 0;           /* cycle counter */
    int count = 0;           /* utility counter */
    int icnt = 0;           /* integration cycle counter */
    int mcnt = 0;           /* median counter */
    int ecnt = 0;           /* envelope counter */
    int env = 0;            /* envelope scan pointer */
    int envbot = 0;         /* first cycle in envelope scan */
    int envtop = NENV-1;    /* last cycle in envelope scan */
    int i, j, temp;         /* utility temps */
    char codesw = 0;        /* gri a/b switch */
    char pllsw = 0;         /* enable pll switch */
    double dtemp, etemp, ftemp, gtemp; /* utility doubles */
    double fmax, fmin;      /* utility max/min values */
    char msg[80] = "\0";    /* status message */
    int tmp[MMAX];          /* int temporary list */
    double ftmp[MMAX];      /* double temporary list */

/*
 * Decode command-line arguments
 */
 * usage: <program name> <gri> <codes> <agc> <vco>
 * <gri> assigned LORAN-C group repetition interval (default 9960)
 * <codes> m for master, s for slaves (first one found) (default m)
 * <agc> initial agc dac (0-255) (default AGC parameter)
 * <vco> initial vco dac (0-255) (default VCO parameter)
 */
    if (argc > 1)
        sscanf(argv[1], "%i", &gri);
    if (argc > 2)
        pulse = *argv[2];
    if (argc > 3)

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        sscanf(argv[3], "%lf", &agcdac);
    if (argc > 4)
        sscanf(argv[4], "%lf", &vcodac);
/*
* Initialization
*
* This section runs only once. It resets the timing generator,
* loads its registers with default values and clears arrays. The
* program then simulates a "l" keystroke and sets the receiver gain
* at minimum to begin receiver calibration.
*/
    outp(TGC, RESET); outp(TGC, LOAD+0x1f); /* reset STC chip */
    outp(TGC, LOADDP+MASTER); outp(TGD, 0xf0); outp(TGD, 0x8a);
    outp(TGC, LOADDP+1);
    for (i = 0; i < 5*3; i++) {
        outp(TGD, init[i]); outp(TGD, init[i]>>8);
    }
    outp(TGC, LOADARM+0x1f); /* let the good times (mode 1) roll */
    sprintf(report, "Calibrating receiver");

/*
* Main loop
*
* This is the main receiver loop and runs until escaped by a ^C
* signal. The main loop runs twice per frame or once each gri
* (pulse groups a and b) and performs the main receiver update
* between the end of pulse group b and the beginning of pulse group
* a. While most program functions are completed in one frame, some
* may persist indefinitely until canceled by another keystroke or
* automatically by the program.
*/
while (1) {

    /*
    * Scan for keyboard functions
    *
    * This section tests for keyboard commands and decodes
    * keystrokes.
    */
    if (kbhit() != 0) {
        kbd = getch();
        switch (kbd) {

            /*
            * The following commands control the phase of the
            * receiver frame relative to the received signal. These
            * are normally needed only for manual signal
            * acquisition.
            */
            case '+': /* shift frequency offset +step/gri */
                freq += step; break;

            case '-': /* shift frequency offset -step/gri */
                freq -= step; break;

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case '0': /* shift frequency to zero offset */
    freq = 0; break;

case ']': /* shift phase +10 us*step */
    phase = step; break;

case '[': /* shift phase -10 us*step */
    phase = -step; break;

/*
 * The following commands adjust various receiver vco
 * and agc bias values. The exact values are determined
 * at initial receiver alignment and compiled in the
 * source code and normally need not be changed in
 * regular operation.
 */
case '}': /* adjust program gain up */
    dervel *= 1.1; break;

case '{': /* adjust program gain down */
    dervel /= 1.1; break;

case ')': /* adjust receiver gain up */
    agcdac++; agc++; break;

case '(': /* adjust receiver gain down */
    agcdac--; agc--; break;

case '>': /* adjust vco bias up */
    vcodac++; break;

case '<': /* adjust vco bias down */
    vcodac--; break;

/*
 * The following commands select which pulse codes are
 * used and determine which set (a or b) to use. These
 * are normally needed only for manual signal
 * acquisition.
 */
case 'm': /* use master pulse codes */
    pulse = 'm'; break;

case 's': /* use slave pulse codes */
    pulse = 's'; break;

case 'x': /* flip pulse code a/b */
    codesw = !codesw; break;

/*
 * The following commands select the receiver gate and
 * integrator gain. These are normally needed only for
 * manual signal acquisition.
 */
case 'u': /* switch to ungated mode */

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        par = 0x02; step = 10; pllsw = 0; break;

case 'g': /* switch to gri mode */
    par = 0x06; step = 10; pllsw = 0; break;

case 'p': /* switch to pci mode */
    par = 0x0a; step = 10; pllsw = 0; break;

case 'e': /* switch to stb mode */
    par = 0x0c; step = 1; pllsw = 0; break;

case 'l': /* open loop (for alignment) */
    pllsw = 1; break;

/*
 * The following commands establish the receiver mode.
 * Normally, the acquisition process sequences modes
 * automatically through four modes in the order below.
 * These commands can be used to restart the process at
 * any point.
 */
case '1': /* calibrate min gain */
    mode = kbd; par = 0x0a; break;

case '2': /* calibrate max gain */
    mode = kbd; par = 0x0a; break;

case '3': /* calibrate normal gain */
    mode = kbd; par = 0x0a; break;

case '4': /* search for pulse-group phase */
    mode = kbd; par = 0x0a; break;

case '5': /* search for envelope phase */
    mode = kbd; par = 0x0c; break;

/*
 * Display receiver status (debug). Note that the
 * display may take longer than a gri, so that the
 * receiver can loose synchronization. Usually,
 * synchronization can be resynchronized simply by
 * flipping the code phase ("x" command).
 */
case 'q':
    outp(TGC, LOADDP+MASTER);
    printf("status %02x  master mode %02x %02x",
        inp(TGC), inp(TGD), inp(TGD));
    outp(TGC, LOADDP+1);
    printf("counters\n");
    for (i = 1; i < 6; i++) {
        printf("%2i", i);
        for (j=1; j<7; j++) printf(" %02x",
            inp(TGD));
        printf("\n");
    }
    printf("agcdac %4.1lf  vcodac %4.1lf\n",

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                agcdac, vcodac);
        break;
    }
}

/*
 * Wait for next gri and accumulate i, q, agc
 *
 * This section first enables automatic adc start at the end of
 * the next gri. It then reads the i integrator (adc channel 0),
 * q integrator (adc channel 1) and agc (adc channel 2). These
 * values are summed for the a and b pulse groups and processed
 * at the end of the b pulse group.
 */
outp(PAR, ENG | par); outp(ADC, 0);          /* i */
while ((inp(ADCGO)&DONE) == 0);
isig += inp(ADC);
outp(PAR, par);
outp(ADC, 1); outp(ADCGO, START);          /* q */
while ((inp(ADCGO)&DONE) == 0);
qsig += inp(ADC);
outp(ADC, 2); outp(ADCGO, START);          /* agc */
while ((inp(ADCGO)&DONE) == 0);
agcraw += inp(ADC);

/*
 * Process i-phase, q-phase and agc
 *
 * Note that a LORAN frame consists of two gri intervals a and
 * b, each with individual pulse codes. The receiver integrates
 * each gri using the assigned pulse codes. There are two sets
 * of pulse codes, one for the master station and the other for
 * slave stations, of which there may be as many as four. Each
 * LORAN chain is assigned a unique gri interval in the range
 * 40-100 ms.
 */
if (codesw == 0) {
    count++;

    /*
     * gri a processing
     *
     * This section processes the i, q and agc samples from the
     * previous frame and computes the receiver vco and agc dac
     * values. It begins by computing the average in-phase,
     * quadrature-phase and agc signals. Note the offset and
     * sign corrections, since the adc operates in unsigned,
     * inverted mode.
     */
    isig = -(isig-iofs)*gain; qsig = -(qsig-qofs)*gain;
    agcavg += (agcraw-agcofs)/AGCFAC;
    if (agcavg < 0) agcavg = 0;

    /*
     * In calibrate mode "1" the receiver gain is set to

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* minimum. The program waits for the averages to settle
* and then calculates the integrator offsets and agc
* minimum value, which takes a few seconds. When done, the
* program simulates a '2' keystroke to complete receiver
* calibration.
*
* In this mode the vco dac is clamped and the agc dac is
* set to minimum (0).
*/
if (mode == '1') {
    vco = vcodac; agc = 0; env = 0;
    iofs -= isig/AGCFAC; qofs -= qsig/AGCFAC;
    dtemp = agcraw-agcofs; agcofs += dtemp/AGCFAC;
    if (isig*isig+qsig*qsig+dtemp*dtemp < 3 &&
        count > AGCFAC*2) {
        mode = '2';          /* continue in mode '2' */
        count = 0;
        sprintf(report, "iofs %4.1lf  qofs %4.1lf  agcofs %4.1lf  gain %4.1lf\n",
            iofs, qofs, agcofs, gain);
    }
}

/*
* In calibrate mode 2 the receiver gain is set to maximum.
* The program waits for the averages to settle and then
* calculates the agc maximum value, which takes a few
* seconds. When done, the program simulates a '3' keystroke
* to establish the normal receiver gain configuration.
*
* In this mode the vco dac is clamped and the agc dac is
* set to maximum (255).
*/
else if (mode == '2') {
    vco = vcodac; agc = 255;
    dtemp = agcraw-agcmax; agcmax += dtemp/AGCFAC;
    if (dtemp*dtemp < 1 && count > AGCFAC*2) {
        mode = '3';          /* continue in mode 3 */
        count = 0; agcavg = 0;
    }
}

/*
* In calibrate mode 3 the receiver gain is determined by
* the receiver agc, which is a peak-indicating type
* sensitive to the peak pulse power in the 90-110 kHz
* spectrum. This step, which takes a few seconds, is
* necessary only to insure the receiver operates at the
* best signal/noise ratio and without overload during the
* scanning process. When done, the program simulates a '4'
* keystroke to begin the pulse-group scan.
*
* In this mode the vco dac is clamped and the agc dac is
* computed from the receiver agc smoothed output.
*
*** This part not finished yet ***

```

```

*/
else if (mode == '3') {
    vco = vcodac; agc = agcdac;
    if (count > AGCFAC*2) {
        mode = '4';          /* continue in mode 4 */
        count = 0; fmax = 0; fmin = 0; pgcnt++;
        sprintf(report, "agcmax %4.1lf  agcavg %4.1lf\nSearching for signal\n",
            agcmax, agcavg);
    }
}

/*
* In acquisition mode 4 the radio scans at 100 us per frame
* over the entire frame (2*gri), which takes up to about
* seven minutes. The program accumulates the 12 most recent
* received rms signal samples in a pulse-gate filter. The
* first two samples are used as a noise gate. In the three
* most recent samples, including two in the filter and the
* most recent, either the first or last must be at least
* 1/3 the second, or the second is most likely a noise
* pulse, which could be due to a pulse code from another
* chain just happening to appear in the pulse-code window.
* In addition, the program computes a weighted sum which
* favors the two samples near the middle of the last ten
* stages in the filter. The program saves the value and
* index of the highest sample received. Note that sample
* selection from the pulse-group filter is valid only after
* 12 samples have been received. Therefore, the program
* waits for 12 samples before updating a bin and wraps
* around for 12 bins beyond the end of the frame. In order
* to improve the estimated position, the program
* interpolates between the intervals just before and just
* after the selected interval. When the scan is complete,
* the program simulates a '5' keystroke to begin the
* envelope scan.
*
* In this mode the vco dac is clamped and the agc dac
* remains at the value computed in the previous mode.
*/
else if (mode == '4') {
    vco = vcodac; agc = agcdac;
    if (count > 2*gri/10+NRMS) {
        fmin = sqrt(fmin/count);
        if (fmax <= 4*fmin) {
            status(fmax, fmin, " low signal");
            count = 0; fmax = 0; fmin = 0; pgcnt++;
        }
    }
    else {
        status(fmax, fmin, " cycle search");
        if (fmin > 0)
            psnr = fmax/fmin;
        count = 0; pllsw = 0;
        phase = mindex-offset-OFFSET; freq = 0;
        mode = '5';          /* continue in mode 5 */
        envbot = 0; envtop = NENV-1;
    }
}

```

```

        env = 0; ptrenv = 0; par = 0x0c;
        for (i = 0; i < MMAX; i++)
            stb[i] = 0;
        efac = EMIN; pfac = PMIN;
        esnr = 0; mcnt = 0; ecnt = 0;
        strobe = 0; fmin = 999; fmax = 0;
        icnt = 0; encnt++;
    }
}
else {
    freq = 10;
    dtemp = isig*isig+qsig*qsig;
    fmin += dtemp;
    dtemp = sqrt(dtemp);
    if (rms[0] > dtemp*PGATE ||
        rms[0] > rms[1]*PGATE) {
        rms[0] = (dtemp+rms[1])/2; pncnt++;
    }
    for (i = 0; i < NRMS-1; i++) {
        rms[i+1] = rms[i];
    }
    rms[0] = dtemp;
    dtemp = rms[4]+rms[5]+rms[6];
    dtemp -= (rms[1]+rms[2]+rms[3]+
        rms[7]+rms[8]+rms[9])/6;
    if (count > NRMS) {
        if (dtemp > fmax) {
            fmax = dtemp;
            dtemp = (rms[6]-rms[4])/
                (rms[6]+rms[4]);
            mindex = offset+(int)dtemp*10;
        }
    }
    if (((count+1)%100) == 0)
        status(fmax, sqrt(fmin/count), "\0");
}
}

/*
 * In tracking mode 5, which is the default, the program
 * scans at 10 us per frame over the 300-us receiver
 * pulse-gate interval. The program integrates the q-
 * channel and i-channel envelopes separately in one-cycle
 * (10 us) bins and determines the reference cycle as the
 * minimum rms error relative to a prestored model envelope.
 * If sufficient integrated signal amplitude is not found
 * within a minute or so, the search is abandoned and the
 * program reverts to the pulse-group search.
 *
 * In this mode the vco dac follows the i channel and the
 * agc follows the q channel, both averaged in complicated
 * ways.
 */
else {

```

```

/*
 * This is the envelope scan. The q signal represents
 * the amplitude and the i signal the phase-correction.
 * The program actually uses only the envelope
 * amplitude.
 */
ftemp = sqrt(isig*isig+qsig*qsig);
gtemp = isig;

/*
 * Median filters for both the amplitude and phase
 * signals for each cycle of the pulse are used to help
 * cope with cross-rate interference and dual-rate
 * transmitter blanking. The variables pmed and amed are
 * shift registers containing the most recent samples.
 * The variables ftemp and gtemp are the median
 * amplitude and phase channels, respectively, while the
 * variable mspan is the span of the envelope channel
 * for later use as a noise gate.
 */
if (mcnt < MMIN)
    mcnt++;
for (i = 0; i < mcnt-1; i++) {
    pmed[env][i+1] = pmed[env][i];
    emed[env][i+1] = emed[env][i];
}
pmed[env][0] = ftemp;
emed[env][0] = gtemp;
for (i = 0; i < mcnt; i++){
    ftmp[i] = pmed[env][i];
    for (j = 0; j < i; j++) {
        if (ftmp[i] < ftmp[j]) {
            gtemp = ftmp[i]; ftmp[i] = ftmp[j];
            ftmp[j] = gtemp;
        }
    }
}
ftemp = ftmp[mcnt/2];
mgate = ftmp[mcnt-1]-ftmp[0];
for (i = 0; i < mcnt; i++){
    ftmp[i] = emed[env][i];
    for (j = 0; j < i; j++) {
        if (ftmp[i] < ftmp[j]) {
            gtemp = ftmp[i]; ftmp[i] = ftmp[j];
            ftmp[j] = gtemp;
        }
    }
}
gtemp = ftmp[mcnt/2];
ecyc[env] += (ftemp-ecyc[env])/efac;
pcyc[env] += (gtemp-pcyc[env])/pfac;

/*
 * Experiment: compute derived envelope.
 */

```

```

if (env == envtop)
    ftemp = ecyc[env];
else
    ftemp = ecyc[env+1];
edrv[env] = ecyc[env]-dervel*(ftemp-ecyc[env]);
if (edrv[env] < 0)
    edrv[env] = -edrv[env];

/*
 * Compute rms envelope error and find cycles of max
 * amplitude and min error.
 */
if (ecyc[env] > fmax) {
    fmax = ecyc[env]; maxdex = env;
}
if (env+6 > NENV-1)
    ftemp = ecyc[NENV-1];
else
    ftemp = ecyc[env+6];
if (ftemp != 0)
    ftemp = 100/ftemp;
etemp = 0;
for (i = 0; i < 7; i++) {
    j = env+i;
    if (j > envtop)
        dtemp = ecyc[envtop]*ftemp;
    else
        dtemp = ecyc[j]*ftemp;
    dtemp -= envcyc[i]; etemp += dtemp*dtemp;
}
erms[env] = sqrt(etemp/7);
if (erms[env] < fmin) {
    fmin = erms[env]; mindex = env;
}

/*
if (edrv[env] < fmin &&
    env < maxdex-1 && env > maxdex-7) {
    fmin = edrv[env]; mindex = env-2;
}

*/

/*
 * This section is entered at the end of the envelope
 * scan. It establishes the cycle position and
 * calculates the signal/noise ratio and phase-
 * correction signal and receiver gain control. The
 * cycle position is determined from the envelope cycle
 * of minimum rms error using a median filter. The
 * phase-correction signal is extracted from the third
 * carrier cycle, while the signal/noise ratio is
 * computed as the ratio of the amplitude of the seventh
 * carrier cycle divided by the rms error of the first
 * cycle. The span of the values in the strobe filter is
 * calculated for later use as a noise gate. In order to
 * maintain critical damping (damping factor of 0.707),

```

```

* the pll gain must be controlled so that the product
* of the loop time constant and loop gain is equal to
* 1/2.
*/
if (env == envbot) {
    fmin = 999; fmax = 0;
    for (i = 0; i < mcnt-1; i++)
        stb[i+1] = stb[i];
    if (maxdex < 8)
        stb[0] = 0;
    else if (mindex < maxdex-8 || mindex > maxdex-4)
        stb[0] = maxdex-6;
    else
        stb[0] = mindex;
    for (i = 0; i < mcnt; i++){
        tmp[i] = stb[i];
        for (j = 0; j < i; j++) {
            if (tmp[i] < tmp[j]) {
                temp = tmp[i]; tmp[i] = tmp[j];
                tmp[j] = temp;
            }
        }
    }
    cycle = tmp[mcnt/2];
    ftemp = 1/(2*VGAIN*pfac);
    if (ftemp > 1)
        ftemp = 1;
    ftemp *= pcy[cycle+2];
    if (pllsw != 0) ftemp = 0;
    vco = vcodac+ftemp;
    if (cycle < 1 || cycle > 17)
        sgate = SGATE;
    else
        sgate = tmp[mcnt-1]-tmp[0];
    dtemp = ecyc[cycle+6]; etemp = erms[cycle];
    if (etemp > 0)
        esnr = dtemp/etemp;
    else
        esnr = 0;
    strcpy(msg, "\0");

/*
* This section is entered at the end of each
* integration cycle. It checks the signal quality,
* cycle position within the 300-us envelope gate
* and strobe position within the 90-us cycle-
* identification window. The receiver gain is
* determined from the amplitude of the seventh
* cycle in the window. If reliable cycle
* identification has not been achieved in a
* reasonable time, the program punts back to the
* pulse-group scan. The integration time constant
* is increased only if all samples in the strobe
* median filter are identical.
*/

```

```

icnt++;
if (icnt >= mcnt) {
    icnt = 0;
    if (mcnt < (int)(8*pfac) && mcnt < MMAX)
        mcnt++;
    if (sgate >= SGATE) {
        if (count < WATCHDOG) {
            strcpy(msg, " strobe noise");
            if (strobe != 0) {
                mode = '5'; /* continue in mode 5 */
                envbot = 0; envtop = NENV-1;
                env = 0; ptrenv = 0; par = 0x0c;
                for (i = 0; i < MMAX; i++)
                    stb[i] = 0;
                efac = EMIN; pfac = PMIN;
                esnr = 0; mcnt = 0; ecnt = 0;
                strobe = 0; fmin = 999; fmax = 0;
                encnt++;
            }
        }
        else {
            strcpy (msg,
                " resume pulse-group search");
            mode = '4'; /* continue in mode 4 */
            count = 0; fmax = 0; fmin = 0; pgcnt++;
            par = 0x0a;
        }
    }
    else {
        if (cycle < 3 || cycle >= 15) {
            strcpy(msg, " cycle slip");
            phase = -(9-cycle); cscnt++;
            mode = '5'; /* continue in mode 5 */
            envbot = 0; envtop = NENV-1;
            env = 0; ptrenv = 0; par = 0x0c;
            for (i = 0; i < MMAX; i++)
                stb[i] = 0;
            efac = EMIN; pfac = PMIN;
            esnr = 0; mcnt = 0; ecnt = 0;
            strobe = 0; fmin = 999; fmax = 0;
            cscnt++;
        }
        else if (cycle+2 != strobe) {
            strcpy(msg, " strobe slip");
            mode = '6'; sscnt++;
            strobe = cycle+2;
        }
    }
}
ecnt++;
if (ecnt >= (int)efac) {
    ecnt = 0;
    if (strobe != 0) {
        if (dtemp > 100)
            agc--;
    }
}

```

```

        else
            agc++;
        }
    if (sgate == 0) {
        if (efac < EMAX)
            efac *= EFAC;
        if (pfac < PMAX)
            pfac *= PFAC;
        }
    }

/*
 * This section determines what to display about the
 * status of the receiver and tracking information.
 * The envelope amplitude phase-correction and rms
 * error signals can be displayed for every cycle
 * scanned. In coarse-search mode before the strobe
 * position has been determined only the
 * signal/noise ratio and strobe span are displayed;
 * while, in fine-tracking mode only the three
 * cycles before and five cycles after the strobe
 * are displayed. This includes one cycle before the
 * cycle-position window and one after. This reduces
 * the integrator bumps in case of a strobe slip,
 * which normally is not more than +-1 cycle.
 */
if (strobe == 0) {
    strcat(msg, " ?");
    dtemp = esnr; etemp = sgate;
    ptrenv = cycle+2;
}
else {
    count = 0;
    dtemp = ecy[cycle]; etemp = erms[cycle];
    if (ptrenv == strobe)
        strcat(msg, " x");
    if (kbd == 'v')
        etemp = pcy[cycle];
    if (ptrenv == envtop-1) {
        strcat(msg, " agc");
        etemp = agc;
    }
    else if (ptrenv == envtop) {
        strcat(msg, " vco");
        etemp = vco;
    }
}
status(dtemp, etemp, msg);
if (strobe != 0) {
    envbot = strobe-3; envtop = strobe+5;
}
else {
    envbot = 0; envtop = NENV-1;
}
ptrenv++;

```

```

        if (ptrenv > envtop) ptrenv = envbot;
        if (ptrenv < envbot) ptrenv = envtop;
    }
}

/*
 * Compute the interval to the next gri. Normally, this is
 * fixed at the gri interval specified less 800 (8 ms), but
 * can be adjusted in 10-us steps to align the timing
 * generator to the transmitted LORAN-C signal. The
 * adjustments can be in the form of a frequency offset, in
 * 10 us/frame increments, or a one-time phase adjustment,
 * in 10-us increments. Note that the interval to the next
 * gri has already been loaded in the counters at this
 * point. The value loaded here actually applies to the gri
 * after that. Since this is the end of a b interval and the
 * next a interval has already been established, frequency
 * and phase adjustments will take effect at the beginning
 * of the next b interval after that, so correct timing
 * between the a and b pulse groups for a single frame are
 * preserved. Also, if a phase adjustment occurs, purge the
 * envelope averages.
 */
offset += freq+phase;          /* adjust epoch */
while (offset >= 2*gri) {
    offset -= 2*gri; frame++;
}
while (offset < 0) {
    offset += 2*gri; frame--;
}
temp = gri+freq+phase;        /* gri counter */
while (temp >= 2*gri)
    temp -= 2*gri;
while (temp < FGUARD || (temp < DGUARD &&
    report[0] != '\0'))
    temp += 2*gri;
temp -= 800; outp(TGC, LOADDP+0x0a);
outp(TGD, temp); outp(TGD, temp>>8); phase = 0;

/*
 * Load the vco and agc dacs and initialize the pulse code
 * and internal integrators for the next frame. Also step
 * the envelope strobe one cycle (10 us) for the next frame.
 */
dtemp = vco;                  /* vco dac */
if (dtemp > 255)
    dtemp = 255;
if (dtemp < 0)
    dtemp = 0;
outp(DACA, (int)dtemp);
dtemp = agc;                  /* agc dac */
if (dtemp > 255)
    dtemp = 255;
if (dtemp < 0)
    dtemp = 0;

```

```

    outp(DACB, (int)dtemp);
    temp = 5000-pcx;          /* envelope scan window */
    outp(TGC, LOADDP+0x0b);
    outp(TGD, pcx); outp(TGD, pcx>>8);
    outp(TGD, temp); outp(TGD, temp>>8);
    outp(TGC, LOADDP+0x0c);
    env--;                    /* envelope scan strobe */
    if (env < envbot)
        env = envtop;
    if (env > envtop)
        env = envbot;
    temp = env*50+RCVDELAY;
    outp(TGD, temp); outp(TGD, temp>>8);
    if (pulse == 's')
        outp(CODE, SPCA); /* pulse code */
    else
        outp(CODE, MPCA);
    isig = 0; qsig = 0; agcraw = 0;
}
else {

    /*
    * gri b processing
    *
    * This section sets up for the b pulse group. It resets the
    * gri (counter 2) load register to delay exactly one gri
    * less 800 (8 ms), which will be used for the subsequent a
    * interval following the next b interval. It also sets the
    * b code for the next b interval. The program also displays
    * the report string if left by the preceding a interval.
    */
    temp = gri-800;          /* gri counter */
    outp(TGC, LOADDP+0x0a);
    outp(TGD, temp); outp(TGD, temp>>8);
    if (pulse == 's')
        outp(CODE, SPCB);
    else
        outp(CODE, MPCB);
    if (report[0] != '\0') {
        puts(report); report[0] = '\0';
    }
}
codesw = !codesw;
}

/*
* subroutine to encode and display status line
* displays line of the form
*
* kggggm n ttt ff ooooo cc uu.u vv.v message
*
* kbd assigned gri
* k last keystroke
* gggg assigned gri

```

```

*   m           master (m) or slave (s) indicator
*   n           operating mode number
*   ttt         time constant
*   ff          frame offset relative to turnon
*   ooooo       cycle offset within the frame (10 us steps) to gri
*   cc          cycle offset within the gri (10 us steps)
*   uu.u        signal value at this position
*   vv.v        error value at this position
*   message     information message
*/
void status(val1, val2, string) double val1, val2; char *string; {

    if (kbd == 'r') {
        sprintf(report, "pg%3i en%3i cs%3i ss%3i pn%5i vc%5.1lf es%5.1lf gn%5.2lf",
            pgcnt, encnt, cscnt, sscnt, pncnt, psnr, esnr,
            gain);
        kbd = ' ';
    }
    else
        sprintf(report, "%c%5i%c %c%6.1f%3i%6i%3i%6.1lf%6.1lf%6.1lf%s",
            kbd, gri, pulse, mode, pfac, frame, offset+strobe, ptrenv,
            val1, edrv[ptrenv], val2, string);
}

/* end program */

```